

I) Introduction

Goals of this study

- Computing time-frequency representations that are
- ▶ candidate for a real-time implementation
- ▶ adjustable by the user through a damping parameter μ (Levenberg-Marquardt approach)
- ▶ able to extract modes and to reconstruct the signal (synchrosqueezing)
- ▶ based on a filter bank approach

Proposed approach

- ▶ use of a particular case of the STFT with a causal, infinite length window function that can be rewritten as a causal IIR recursive filtering
- ▶ the algorithmic complexity depends on the filter order and on the analyzed frequency bandwidth

II) Filter-based reassigned and synchrosqueezed STFT

The STFT as a convolution product

The STFT of a signal x using a real-valued analysis window h , denoted $F_x^h(t, \omega) = M_x^h(t, \omega) e^{j\Phi_x^h(t, \omega)}$ can be related to the linear convolution product between the analyzed signal x and the complex valued impulse response of a bandpass filter $g(t, \omega) = h(t) e^{j\omega t}$:

$$y_x^g(t, \omega) = \int_{-\infty}^{+\infty} g(\tau, \omega) x(t - \tau) d\tau = |y_x^g(t, \omega)| e^{j\Psi_x^g(t, \omega)}$$

$$= F_x^h(t, \omega) e^{j\omega t} = M_x^h(t, \omega) e^{j(\Phi_x^h(t, \omega) + \omega t)}$$

thus

$$M_x^h(t, \omega) = |y_x^g(t, \omega)| \text{ and } \Phi_x^h(t, \omega) = \Psi_x^g(t, \omega) - \omega t.$$

Rewording the reassignment operators of the spectrogram

According to [1], the spectrogram reassignment operators can be reformulated using the phase of $y_x^g(t, \omega)$, denoted $\Psi_x^g(t, \omega) = \Phi_x^h(t, \omega) + \omega t$

$$\hat{t}(t, \omega) = -\frac{\partial \Phi_x^h}{\partial \omega}(t, \omega) = t - \frac{\partial \Psi_x^g}{\partial \omega}(t, \omega), \quad (1)$$

$$\hat{\omega}(t, \omega) = \omega + \frac{\partial \Phi_x^h}{\partial t}(t, \omega) = \frac{\partial \Psi_x^g}{\partial t}(t, \omega). \quad (2)$$

The reassigned spectrogram is expressed as

$$RSP(t, \omega) = \iint_{\mathbb{R}^2} |y_x^g(t', \omega')|^2 \delta(t - \hat{t}(t', \omega')) \delta(\omega - \hat{\omega}(t', \omega')) dt' d\omega' \quad (3)$$

where $\delta(t)$ denotes the Dirac distribution.

Rewording the Levenberg-Marquardt reassignment [2]

$$\begin{pmatrix} \hat{t}(t, \omega) \\ \hat{\omega}(t, \omega) \end{pmatrix} = \begin{pmatrix} t \\ \omega \end{pmatrix} - (\nabla^t R_x^h(t, \omega) + \mu I_2)^{-1} R_x^h(t, \omega) \quad (4)$$

$$R_x^h(t, \omega) = \begin{pmatrix} t - \hat{t}(t, \omega) \\ \omega - \hat{\omega}(t, \omega) \end{pmatrix} = \begin{pmatrix} \frac{\partial \Psi_x^g}{\partial \omega}(t, \omega) \\ \frac{\partial \Psi_x^g}{\partial t}(t, \omega) \end{pmatrix}$$

$$\nabla^t R_x^h(t, \omega) = \begin{pmatrix} \frac{\partial^2 \Psi_x^g}{\partial t \partial \omega}(t, \omega) & \frac{\partial^2 \Psi_x^g}{\partial \omega^2}(t, \omega) \\ -\frac{\partial^2 \Psi_x^g}{\partial t^2}(t, \omega) & 1 - \frac{\partial^2 \Psi_x^g}{\partial t \partial \omega}(t, \omega) \end{pmatrix}$$

Hence, the Levenberg-Marquardt reassigned spectrogram (LMRSP(t, ω)) is obtained by replacing $(\hat{t}, \hat{\omega})$ by $(\tilde{t}, \tilde{\omega})$ in Eq. (3).

Rewording the partial derivatives of the phase

$$\frac{\partial \Psi_x^g}{\partial t}(t, \omega) = \text{Im} \left(\frac{y_x^{Dg}(t, \omega)}{y_x^g(t, \omega)} \right)$$

$$\frac{\partial \Psi_x^g}{\partial \omega}(t, \omega) = \text{Re} \left(\frac{y_x^{Tg}(t, \omega)}{y_x^g(t, \omega)} \right)$$

$$\frac{\partial^2 \Psi_x^g}{\partial t \partial \omega}(t, \omega) = \text{Re} \left(\frac{y_x^{DTg}(t, \omega)}{y_x^g(t, \omega)} - \frac{y_x^{Dg}(t, \omega) y_x^{Tg}(t, \omega)}{y_x^g(t, \omega)^2} \right)$$

$$\frac{\partial^2 \Psi_x^g}{\partial t^2}(t, \omega) = \text{Im} \left(\frac{y_x^{D^2g}(t, \omega)}{y_x^g(t, \omega)} - \frac{(y_x^{Dg}(t, \omega))^2}{y_x^g(t, \omega)^2} \right)$$

$$\frac{\partial^2 \Psi_x^g}{\partial \omega^2}(t, \omega) = -\text{Im} \left(\frac{y_x^{T^2g}(t, \omega)}{y_x^g(t, \omega)} - \frac{(y_x^{Tg}(t, \omega))^2}{y_x^g(t, \omega)^2} \right)$$

where $y_x^g, y_x^{Tg}, y_x^{Dg}, y_x^{DTg}, y_x^{T^2g}$ and $y_x^{D^2g}$ are the outputs of the filters using respectively the impulse responses $g(t, \omega), Tg = t g(t, \omega), Dg(t, \omega) = \frac{\partial g}{\partial t}(t, \omega), DTg(t, \omega) = \frac{\partial}{\partial t}(t g(t, \omega)), T^2g(t, \omega) = t^2 g(t, \omega)$ and $D^2g(t, \omega) = \frac{\partial^2 g}{\partial t^2}(t, \omega)$.

Rewording the synchrosqueezed STFT

y_x^g admits the following signal reconstruction formula

$$x(t - t_0) = \frac{1}{h(t_0)} \int_{-\infty}^{+\infty} y_x^g(t, \omega) e^{-j\omega t_0} \frac{d\omega}{2\pi}, \text{ when } h(t_0) \neq 0. \quad (5)$$

This leads to the synchrosqueezed STFT [5]:

$$Sy_x^g(t, \omega) = \int_{\mathbb{R}} y_x^g(t, \omega') e^{-j\omega' t_0} \delta(\omega - \hat{\omega}(t, \omega')) d\omega' \quad (6)$$

LMSy $_x^g(t, \omega)$ is obtained by replacing $\hat{\omega}$ by $\tilde{\omega}$. A sharpen time-frequency representation is provided by $|Sy_x^g(t, \omega)|^2$. The signal can be reconstructed from $Sy_x^g(t, \omega)$ as

$$\hat{x}(t - t_0) = \frac{1}{h(t_0)} \int_{-\infty}^{+\infty} Sy_x^g(t, \omega) \frac{d\omega}{2\pi} \quad (7)$$

III) Towards a recursive implementation

Recursive implementation [3]

y_x^g can be recursively implemented using

$$h_k(t) = \frac{t^{k-1}}{T^k(k-1)!} e^{-t/T} U(t), \quad (8)$$

$$g_k(t, \omega) = h_k(t) e^{j\omega t} = \frac{t^{k-1}}{T^k(k-1)!} e^{pt} U(t) \quad (9)$$

with $p = -\frac{1}{T} + j\omega$, $k \geq 1$ being the filter order, T the time spread of the window and $U(t)$ the Heaviside step function. **The time derivatives of $g_k(t, \omega)$ or its products by t and t^2 can be expressed as a linear combination of $g_k(t, \omega)$ with different filter orders.**

Discretization using the impulse invariance method [4]

$$G_k(z, \omega) = T_s \mathcal{Z} \{g_k(t, \omega)\} = \frac{\sum_{i=0}^{k-1} b_i z^{-i}}{1 + \sum_{i=1}^k a_i z^{-i}} \quad (10)$$

with $b_i = (L^k(k-1)!)^{-1} B_{k-1, k-i-1} \alpha^i$, $\alpha = e^{pT_s}$, $L = T/T_s$, $a_i = A_{k,i} (-\alpha)^i$, T_s being the sampling period. $B_{k,i} = \sum_{j=0}^i (-1)^j A_{k+1, j} (i+1-j)^k$ denotes the Eulerian numbers and $A_{k,i}$ the binomial coefficients. Hence, using $y_k[n, m] \approx y_x^g(nT_s, \frac{2\pi m}{MT_s})$ with $n \in \mathbb{Z}$ and $m = 0, 1, \dots, M-1$, we obtain

$$y_k[n, m] = \sum_{i=0}^{k-1} b_i x[n-i] - \sum_{j=1}^k a_j y_k[n-j, m] \quad (11)$$

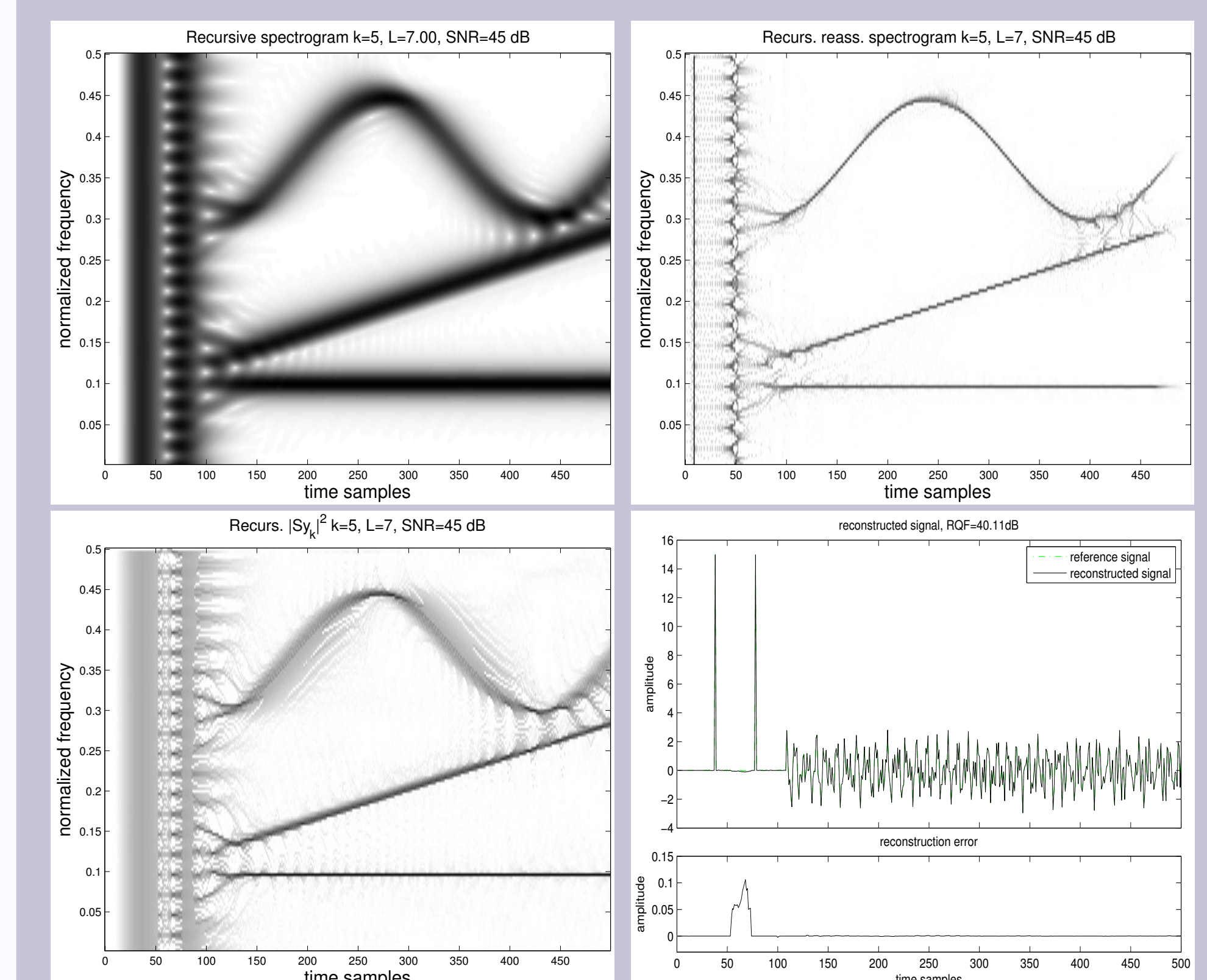
Algorithm implementation for recursive TFR computation

At time index n :

1. Compute the required $y_k^g[n, m]$ using $x[n-i]$ and $y_k[n-j, m]$ with $i \in [0, k-1], j \in [1, k]$
2. Compute the other required specific filtered signals (i.e. $y_k^{Tg}, y_k^{Dg}, y_k^{DTg}, y_k^{T^2g}$ or $y_k^{D^2g}$) using y_k^g with different filter orders
3. Compute the reassignment operators \hat{n}, \hat{m} (resp. \tilde{n}, \tilde{m})
4. **If $\hat{n} \leq n$ (resp. $\tilde{n} \leq n$) then update TFR $[\hat{n}, m]$ else store the triplet $(y_k^g[n, m], \hat{n}, m)$ into a list**
5. Update TFR $[n, m]$ using all previously stored triplets such as $\hat{n} = n$ (resp. $\tilde{n} = n$) and remove them from the list

IV) Numerical results

Resulting time-frequency representations



Signal reconstruction quality

Using the discrete-time version of Eq. (7) with $n_0 = t_0/T_s$.

	n_0	8	18	26	28	30
(a)	RQF (dB)	9.79	24.17	26.77	26.82	26.73
(b)	M	100	200	600	1000	2400
	RQF (dB)	20.56	24.90	29.48	30.50	30.87
(c)	μ	0.30	0.80	1.30	1.80	2.30
	RQF (dB)	20.83	27.28	29.68	30.35	30.90

Signal Reconstruction Quality Factor RQF = $10 \log_{10} \left(\frac{\sum_n |x[n]|^2}{\sum_n |x[n] - \hat{x}[n]|^2} \right)$, of the recursive synchrosqueezed STFT computed for $k = 5, L = 7$ at SNR = 45 dB. Line (a), computed for $M = 300$, Line (b) and Line (c), computed for $n_0 = 28$ and $M = 300$.

Bibliography

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